

Model Order Reduction of Electrical Circuits with Nonlinear Elements

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- ▶ Introduction, Motivation
 - ▶ Model equations for electrical circuits with nonlinear elements
 - ▶ Model order reduction
 - ▶ Software package: PABTEC
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- ▶ Simulation of such large models is mostly impossible or, at least, unacceptably time and storage consuming.
- ▶ **Model order reduction presents a way out of this dilemma.**
- ▶ A general idea of model order reduction is to replace a large-scale system by a much smaller model which approximates the input-output relation of the large-scale system within a required accuracy.

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$$\begin{aligned}\mathcal{E}(x) \frac{d}{dt} x &= \mathcal{A}x + f(x) + \mathcal{B}u, \\ y &= \mathcal{B}^T x,\end{aligned}$$

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$$y = \mathcal{B}^T x,$$

states $x = [\eta^T \quad v_{\mathcal{L}}^T \quad v_{\mathcal{V}}^T]^T$

inputs $u = [v_{\mathcal{I}}^T \quad u_{\mathcal{V}}^T]^T$

outputs $y = - [u_{\mathcal{I}}^T \quad v_{\mathcal{V}}^T]^T$

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$$\mathcal{E}(x) = \begin{bmatrix} A_{\mathcal{C}} C (A_{\mathcal{C}}^T \eta) A_{\mathcal{C}}^T & 0 & 0 \\ 0 & \mathcal{L}(v_{\mathcal{L}}) & 0 \\ 0 & 0 & 0 \end{bmatrix},$$

$$\mathcal{A} = \begin{bmatrix} 0 & -A_{\mathcal{L}} & -A_{\mathcal{V}} \\ A_{\mathcal{C}}^T & 0 & 0 \\ A_{\mathcal{V}}^T & 0 & 0 \end{bmatrix},$$

$$f(x) = \begin{bmatrix} -A_{\mathcal{R}} g(A_{\mathcal{R}}^T \eta) \\ 0 \\ 0 \end{bmatrix},$$

$$\mathcal{B} = \begin{bmatrix} -A_{\mathcal{I}} & 0 \\ 0 & 0 \\ 0 & -I \end{bmatrix}.$$

$$\mathcal{E}(x) \frac{d}{dt} x = \mathcal{A}x + f(x) + \mathcal{B}u,$$

$$y = \mathcal{B}^T x,$$

\mathcal{C} , \mathcal{R} , \mathcal{L} , \mathcal{V} , \mathcal{I} as index denote conductors, resistors, inductors, voltage sources, current sources

η vector of node potentials

i_* vector of currents

u_* vector of voltages

A_* incidence matrices

\mathcal{C} conductance matrix-valued function

\mathcal{L} inductance matrix-valued function

g resistor relation

states $x = [\eta^T \quad i_{\mathcal{L}}^T \quad i_{\mathcal{V}}^T]^T$

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$$\mathcal{E}(x) = \begin{bmatrix} A_{\mathcal{C}} \mathcal{C}(A_{\mathcal{C}}^T \eta) A_{\mathcal{C}}^T & 0 & 0 \\ 0 & \mathcal{L}(i_{\mathcal{L}}) & 0 \\ 0 & 0 & 0 \end{bmatrix},$$

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$$f(x) = \begin{bmatrix} -A_{\mathcal{R}} g(A_{\mathcal{R}}^T \eta) \\ 0 \\ 0 \end{bmatrix},$$

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We will assume that

(A1) the matrix $A_{\mathcal{V}}$ has full column rank,

(A2) the matrix $[A_{\mathcal{C}}, A_{\mathcal{L}}, A_{\mathcal{R}}, A_{\mathcal{V}}]$ has full row rank,

We will assume that

- (A1) the matrix A_V has full column rank,
- (A2) the matrix $[A_C, A_L, A_R, A_V]$ has full row rank,
- (A3) the matrices $\mathcal{C}(A_C^T \eta)$ and $\mathcal{L}(v_L)$ are positive definite for all admissible η and v_L , and
- (A4) the function $g(A_R^T \eta)$ is monotonically increasing for all admissible η .

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Assumptions (A1) and (A2) imply that the circuit does not contain loops of voltage sources and cutsets of current sources, respectively, while assumptions (A3) and (A4) on the capacitance and inductance matrices and the resistor relation mean that all circuit elements do not generate energy.

Furthermore, we assume without loss of generality that the circuit elements are ordered such that

$$A_{\mathcal{C}} = \begin{bmatrix} A_{\bar{\mathcal{C}}} & A_{\tilde{\mathcal{C}}} \end{bmatrix}, \quad A_{\mathcal{L}} = \begin{bmatrix} A_{\bar{\mathcal{L}}} & A_{\tilde{\mathcal{L}}} \end{bmatrix}, \\ A_{\mathcal{R}} = \begin{bmatrix} A_{\bar{\mathcal{R}}} & A_{\tilde{\mathcal{R}}} \end{bmatrix},$$

We also assume that the linear and nonlinear elements are not mutually connected, i.e.,

$$C(A_{\mathcal{C}}^T \eta) = \begin{bmatrix} \bar{C} & 0 \\ 0 & \tilde{C}(A_{\tilde{\mathcal{C}}}^T \eta) \end{bmatrix}, \quad \mathcal{L}(v_{\mathcal{L}}) = \begin{bmatrix} \bar{\mathcal{L}} & 0 \\ 0 & \tilde{\mathcal{L}}(v_{\tilde{\mathcal{L}}}) \end{bmatrix}, \\ g(A_{\mathcal{R}}^T \eta) = \begin{bmatrix} \bar{G} A_{\bar{\mathcal{R}}}^T \eta \\ \tilde{g}(A_{\tilde{\mathcal{R}}}^T \eta) \end{bmatrix},$$

Consequently, we have the model equations in the form

$$\begin{aligned}\mathcal{E}(x) \frac{d}{dt} x &= \mathcal{A}x + f(x) + \mathcal{B}u, \\ y &= \mathcal{B}^T x,\end{aligned}$$

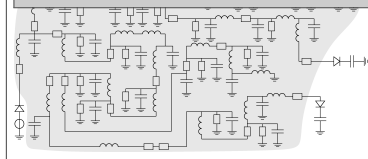
with

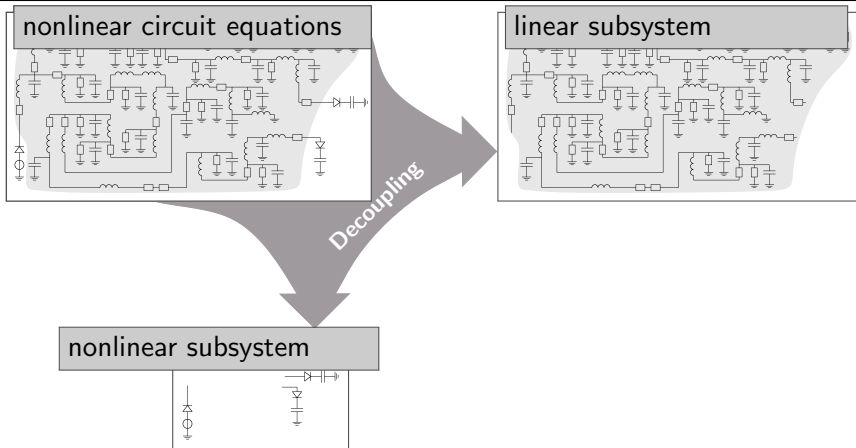
$$\mathcal{E}(x) = \begin{bmatrix} A_{\tilde{c}} \bar{c} A_{\tilde{c}}^T + A_{\tilde{c}} \tilde{c} (A_{\tilde{c}}^T \eta) A_{\tilde{c}}^T & 0 & 0 & 0 \\ 0 & \tilde{\mathcal{L}} & 0 & 0 \\ 0 & 0 & \tilde{\mathcal{L}}(v_{\tilde{\mathcal{L}}}) & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad f(x) = \begin{bmatrix} -A_{\tilde{\mathcal{R}}} \tilde{g}(A_{\tilde{\mathcal{R}}}^T \eta) \\ 0 \\ 0 \\ 0 \end{bmatrix},$$

$$\mathcal{A} = \begin{bmatrix} A_{\tilde{\mathcal{R}}} \tilde{g} A_{\tilde{\mathcal{R}}}^T & -A_{\tilde{\mathcal{L}}} & -A_{\tilde{\mathcal{L}}} & -A_{\mathcal{V}} \\ A_{\tilde{\mathcal{L}}}^T & 0 & 0 & 0 \\ A_{\tilde{\mathcal{L}}}^T & 0 & 0 & 0 \\ A_{\mathcal{V}}^T & 0 & 0 & 0 \end{bmatrix}, \quad \mathcal{B} = \begin{bmatrix} -A_{\mathcal{I}} & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & -I \end{bmatrix}.$$

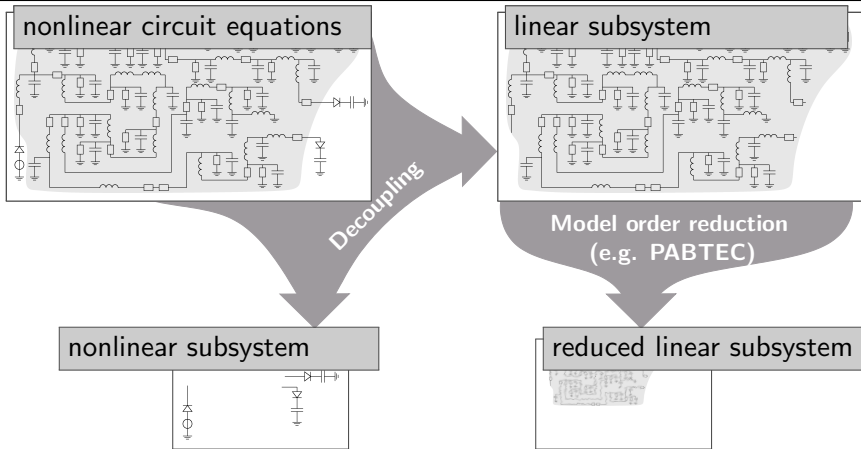
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nonlinear circuit equations

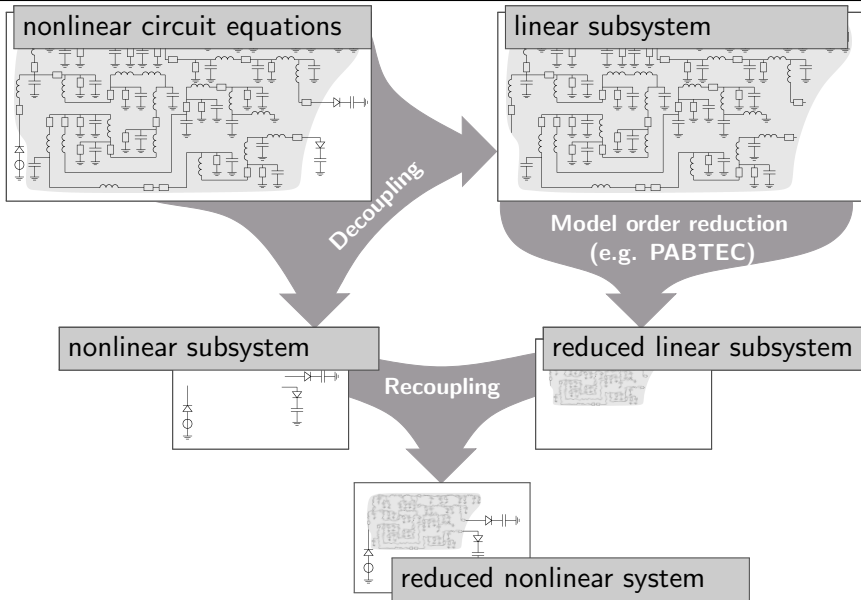


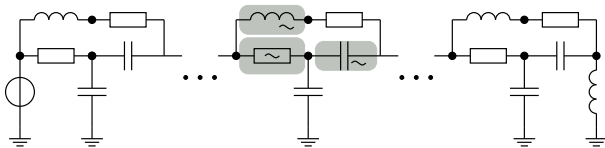


Model Order Reduction - Approach



Model Order Reduction - Approach





:-(possibly creation of
LI-cutsets

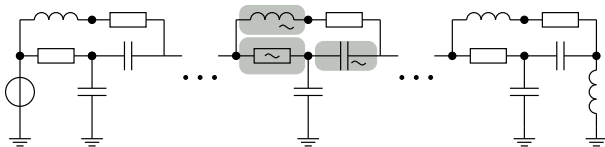


:-)



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Model Order Reduction - Decoupling



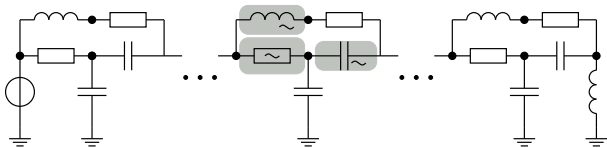
:-/ introduction of additional variables



:-)



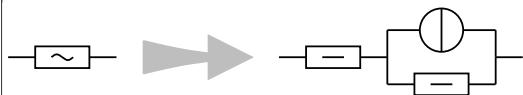
:-(possibly creation of CV-loops



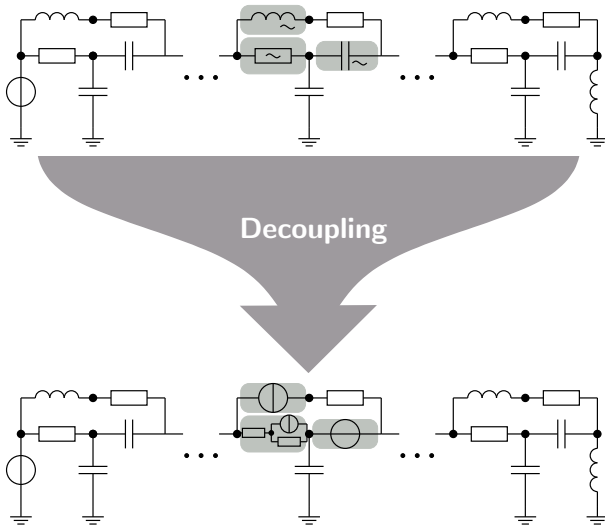
:-/ introduction of additional variables



:-)



:-/ introduction of additional nodes



Let $A_{\tilde{\mathcal{R}}} \in \{-1, 0, 1\}^{n_\eta, n_{\tilde{\mathcal{R}}}}$ be an incidence matrix. Then the matrices $A_{\tilde{\mathcal{R}}}^1$ and $A_{\tilde{\mathcal{R}}}^2$ are uniquely defined with $A_{\tilde{\mathcal{R}}}^1 \in \{0, 1\}^{n_\eta, n_{\tilde{\mathcal{R}}}}$ and $A_{\tilde{\mathcal{R}}}^2 \in \{-1, 0\}^{n_\eta, n_{\tilde{\mathcal{R}}}}$ satisfying $A_{\tilde{\mathcal{R}}}^1 + A_{\tilde{\mathcal{R}}}^2 = A_{\tilde{\mathcal{R}}}$. Furthermore, let $v_{\tilde{\mathcal{L}}} \in \mathbb{R}^{n_{\tilde{\mathcal{L}}}}$, $u_{\tilde{\mathcal{C}}} \in \mathbb{R}^{n_{\tilde{\mathcal{C}}}}$, and $v_z \in \mathbb{R}^{n_{\tilde{\mathcal{R}}}}$ be defined by the relations

$$\tilde{\mathcal{L}}(v_{\tilde{\mathcal{L}}}) \frac{d}{dt} v_{\tilde{\mathcal{L}}} = A_{\tilde{\mathcal{L}}}^T \eta, \quad (1)$$

$$u_{\tilde{\mathcal{C}}} = A_{\tilde{\mathcal{C}}}^T \eta, \quad (2)$$

$$v_z = \Gamma_s G_1^{-1} (\tilde{g}(A_{\tilde{\mathcal{R}}}^T \eta) - \Gamma_{12} A_{\tilde{\mathcal{R}}}^T \eta) \quad (3)$$

with the notation

$$\begin{aligned} \Gamma_s &= G_1 + G_2, \\ \Gamma_{12} &= G_1(G_1 + G_2)^{-1} G_2. \end{aligned}$$

Model Order Reduction - Decoupling

Then the original system of model equations together with the relations

$$\eta_z = \Gamma_s^{-1}((G_1(A_{\tilde{\mathcal{R}}}^1)^T - G_2(A_{\tilde{\mathcal{R}}}^2)^T)\eta - v_z), \quad (4a)$$

$$v_{\tilde{\mathcal{C}}} = \tilde{\mathcal{C}}(u_{\tilde{\mathcal{C}}}) \frac{d}{dt} u_{\tilde{\mathcal{C}}} \quad (4b)$$

is equivalent to the **linear system**

$$\begin{bmatrix} A_{\tilde{\mathcal{C}}}\tilde{\mathcal{C}}A_{\tilde{\mathcal{C}}}^T & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & \tilde{\mathcal{L}} & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \frac{d}{dt}\eta \\ \frac{d}{dt}\eta_z \\ \frac{d}{dt}v_{\tilde{\mathcal{L}}} \\ \frac{d}{dt}v_{\mathcal{V}} \\ \frac{d}{dt}v_{\tilde{\mathcal{C}}} \end{bmatrix} = \begin{bmatrix} \mathcal{A}_{11} & \mathcal{A}_{12} & -A_{\tilde{\mathcal{L}}} & -A_{\mathcal{V}} & -A_{\tilde{\mathcal{C}}} \\ \mathcal{A}_{12}^T & -\Gamma_s & 0 & 0 & 0 \\ A_{\tilde{\mathcal{L}}}^T & 0 & 0 & 0 & 0 \\ A_{\mathcal{V}}^T & 0 & 0 & 0 & 0 \\ A_{\tilde{\mathcal{C}}}^T & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \eta \\ \eta_z \\ v_{\tilde{\mathcal{L}}} \\ v_{\mathcal{V}} \\ v_{\tilde{\mathcal{C}}} \end{bmatrix} \quad (5a)$$

$$+ \begin{bmatrix} -A_{\mathcal{I}} & -A_{\tilde{\mathcal{R}}}^2 & -A_{\tilde{\mathcal{L}}} & 0 & 0 \\ 0 & -I & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & -I & 0 \\ 0 & 0 & 0 & 0 & -I \end{bmatrix} \begin{bmatrix} v_{\mathcal{I}} \\ v_z \\ v_{\tilde{\mathcal{L}}} \\ u_{\mathcal{V}} \\ u_{\tilde{\mathcal{C}}} \end{bmatrix},$$

$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \\ y_4 \\ y_5 \end{bmatrix} = \begin{bmatrix} -A_{\mathcal{I}}^T & 0 & 0 & 0 & 0 \\ -(A_{\tilde{\mathcal{R}}}^2)^T & -I & 0 & 0 & 0 \\ -A_{\tilde{\mathcal{L}}}^T & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & -I & 0 \\ 0 & 0 & 0 & 0 & -I \end{bmatrix} \begin{bmatrix} \eta \\ \eta_z \\ v_{\tilde{\mathcal{L}}} \\ v_{\mathcal{V}} \\ v_{\tilde{\mathcal{C}}} \end{bmatrix} \quad (5b)$$

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$$\begin{bmatrix} y_1 \\ y_2 \\ y_3 \\ y_4 \\ y_5 \end{bmatrix} = \begin{bmatrix} -A_{\mathcal{I}}^T & 0 & 0 & 0 & 0 \\ -(A_{\tilde{\mathcal{R}}}^2)^T & -I & 0 & 0 & 0 \\ -A_{\tilde{\mathcal{L}}}^T & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & -I & 0 \\ 0 & 0 & 0 & 0 & -I \end{bmatrix} \begin{bmatrix} \eta \\ \eta_z \\ v_{\tilde{\mathcal{L}}} \\ v_{\mathcal{V}} \\ v_{\tilde{\mathcal{C}}} \end{bmatrix}$$

with

$$\begin{aligned} \mathcal{A}_{11} &= -A_{\tilde{\mathcal{R}}}G A_{\tilde{\mathcal{R}}}^T - A_{\tilde{\mathcal{R}}}^1 G_1 (A_{\tilde{\mathcal{R}}}^1)^T \\ &\quad - A_{\tilde{\mathcal{R}}}^2 G_2 (A_{\tilde{\mathcal{R}}}^2)^T, \\ \mathcal{A}_{12} &= A_{\tilde{\mathcal{R}}}^1 G_1^T - A_{\tilde{\mathcal{R}}}^2 (G_2)^T \end{aligned}$$

Model Order Reduction - Decoupling

Then the original system of model equations together with the relations

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Keep in mind

$$y_2 = -(A_{\tilde{\mathcal{R}}}^2)^T \eta - \eta_z$$

$$y_3 = -A_{\tilde{\mathcal{L}}}^T \eta$$

$$y_5 = -v_{\tilde{\mathcal{C}}}$$

Application of a model order reduction method like PABTECL yields the reduced-order model

$$\hat{\mathcal{E}} \frac{d}{dt} \hat{x} = \hat{A} \hat{x} + [\hat{B}_1 \hat{B}_2 \hat{B}_3 \hat{B}_4 \hat{B}_5] \begin{bmatrix} v_I \\ v_z \\ v_{\tilde{\mathcal{L}}} \\ u_V \\ u_{\tilde{\mathcal{C}}} \end{bmatrix}, \quad (6a)$$

$$\hat{y} = \begin{bmatrix} \hat{C}_1 \\ \hat{C}_2 \\ \hat{C}_3 \\ \hat{C}_4 \\ \hat{C}_5 \end{bmatrix} \hat{x}. \quad (6b)$$

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Application of a model order reduction method like PABTECL yields the reduced-order model

$$\hat{\mathcal{E}} \frac{d}{dt} \hat{x} = \hat{A} \hat{x} + [\hat{B}_1 \hat{B}_2 \hat{B}_3 \hat{B}_4 \hat{B}_5] \begin{bmatrix} v_I \\ v_z \\ v_{\tilde{\mathcal{L}}} \\ u_V \\ u_{\tilde{\mathcal{C}}} \end{bmatrix},$$

$$\begin{bmatrix} -(A_{\tilde{\mathcal{R}}}^2)^T \eta - \eta_z \\ -A_{\tilde{\mathcal{L}}}^T \eta \\ -v_{\tilde{\mathcal{C}}} \end{bmatrix} \approx \begin{bmatrix} \hat{\mathcal{C}}_1 \\ \hat{\mathcal{C}}_2 \\ \hat{\mathcal{C}}_3 \\ \hat{\mathcal{C}}_4 \\ \hat{\mathcal{C}}_5 \end{bmatrix} \hat{x}.$$

Keep in mind

$$y_2 = -(A_{\tilde{\mathcal{R}}}^2)^T \eta - \eta_z$$

$$y_3 = -A_{\tilde{\mathcal{L}}}^T \eta$$

$$y_5 = -v_{\tilde{\mathcal{C}}}$$

We have
$$-(A_{\tilde{\mathcal{R}}}^2)^T \eta - \eta_z \approx \hat{C}_2 \hat{x}, \quad (12a)$$

$$-A_{\tilde{\mathcal{L}}}^T \eta \approx \hat{C}_3 \hat{x}, \quad (12b)$$

$$-v_{\tilde{\mathcal{C}}} \approx \hat{C}_5 \hat{x}. \quad (12c)$$

Then we get from (1), (3), (4a), and (4b) the relations

$$\tilde{\mathcal{L}}(\hat{v}_{\tilde{\mathcal{L}}}) \frac{d}{dt} \hat{v}_{\tilde{\mathcal{L}}} = -\hat{C}_3 \hat{x} \quad (13a)$$

$$\tilde{\mathcal{C}}(\hat{u}_{\tilde{\mathcal{C}}}) \frac{d}{dt} \hat{u}_{\tilde{\mathcal{C}}} = -\hat{C}_5 \hat{x}, \quad (13b)$$

$$0 = -G_1 \hat{C}_2 \hat{x} - G_1 \hat{u}_{\tilde{\mathcal{R}}} + \tilde{g}(\hat{u}_{\tilde{\mathcal{R}}}) \quad (13c)$$

and

$$v_z = \Gamma_s G_1^{-1} \tilde{g}(u_{\tilde{\mathcal{R}}}) - G_2 u_{\tilde{\mathcal{R}}}, \quad (14)$$

where $\hat{v}_{\tilde{\mathcal{L}}}$, $\hat{u}_{\tilde{\mathcal{C}}}$, and $\hat{u}_{\tilde{\mathcal{R}}}$ are approximations for $v_{\tilde{\mathcal{L}}}$, $u_{\tilde{\mathcal{C}}}$, and $u_{\tilde{\mathcal{R}}}$, respectively.

Now, adding (13a), (13b), and (13c) to (6) and using in addition to \hat{x} also the approximations $\hat{i}_{\tilde{\mathcal{L}}}$, $\hat{u}_{\tilde{\mathcal{C}}}$, and $\hat{u}_{\tilde{\mathcal{R}}}$ as state variables, then we get with (14) the descriptor system

$$\begin{bmatrix} \hat{\mathcal{E}} & 0 & 0 & 0 \\ 0 & \tilde{\mathcal{L}}(\hat{i}_{\tilde{\mathcal{L}}}) & 0 & 0 \\ 0 & 0 & \tilde{\mathcal{C}}(\hat{u}_{\tilde{\mathcal{C}}}) & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \frac{d}{dt} \hat{x} \\ \frac{d}{dt} \hat{i}_{\tilde{\mathcal{L}}} \\ \frac{d}{dt} \hat{u}_{\tilde{\mathcal{C}}} \\ \frac{d}{dt} \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix} = \begin{bmatrix} \hat{A} & \hat{B}_3 & \hat{B}_5 & -\hat{B}_2 G_2 \\ -\hat{C}_3 & 0 & 0 & 0 \\ -\hat{C}_5 & 0 & 0 & 0 \\ -G_1 \hat{C}_2 & 0 & 0 & -G_1 \end{bmatrix} \begin{bmatrix} \hat{x} \\ \hat{i}_{\tilde{\mathcal{L}}} \\ \hat{u}_{\tilde{\mathcal{C}}} \\ \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix} \\ + \begin{bmatrix} \hat{B}_1 & \hat{B}_4 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} v_{\mathcal{I}} \\ u_{\mathcal{V}} \end{bmatrix} + \begin{bmatrix} \hat{B}_2 \Gamma_s G_1^{-1} \tilde{g}(\hat{u}_{\tilde{\mathcal{R}}}) \\ 0 \\ 0 \\ \tilde{g}(\hat{u}_{\tilde{\mathcal{R}}}) \end{bmatrix},$$

$$\begin{bmatrix} \hat{y}_1 \\ \hat{y}_4 \end{bmatrix} = \begin{bmatrix} \hat{C}_1 & 0 & 0 & 0 \\ \hat{C}_4 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{x} \\ \hat{i}_{\tilde{\mathcal{L}}} \\ \hat{u}_{\tilde{\mathcal{C}}} \\ \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix},$$

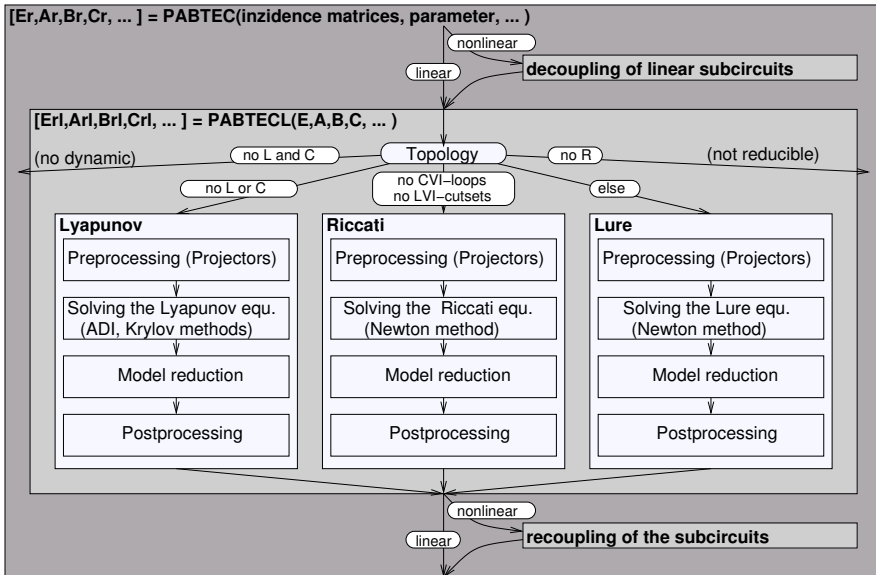
With row manipulations of the state equations we obtain, finally, the nonlinear descriptor system

$$\begin{bmatrix} \hat{\mathcal{E}} & 0 & 0 & 0 \\ 0 & \tilde{\mathcal{L}}(\hat{i}_{\tilde{\mathcal{L}}}) & 0 & 0 \\ 0 & 0 & \tilde{\mathcal{C}}(\hat{u}_{\tilde{\mathcal{C}}}) & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \frac{d}{dt} \hat{x} \\ \frac{d}{dt} \hat{i}_{\tilde{\mathcal{L}}} \\ \frac{d}{dt} \hat{u}_{\tilde{\mathcal{C}}} \\ \frac{d}{dt} \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix} = \begin{bmatrix} \hat{A} + \hat{B}_2 \Gamma_s \hat{C}_2 & \hat{B}_3 & \hat{B}_5 & \hat{B}_2 G_1 \\ -\hat{C}_3 & 0 & 0 & 0 \\ -\hat{C}_5 & 0 & 0 & 0 \\ -G_1 \hat{C}_2 & 0 & 0 & -G_1 \end{bmatrix} \begin{bmatrix} \hat{x} \\ \hat{i}_{\tilde{\mathcal{L}}} \\ \hat{u}_{\tilde{\mathcal{C}}} \\ \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix} \\ + \begin{bmatrix} \hat{B}_1 & \hat{B}_4 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} v_{\mathcal{I}} \\ u_{\mathcal{V}} \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 0 \\ \tilde{g}(\hat{u}_{\tilde{\mathcal{R}}}) \end{bmatrix},$$

$$\begin{bmatrix} \hat{y}_1 \\ \hat{y}_4 \end{bmatrix} = \begin{bmatrix} \hat{C}_1 & 0 & 0 & 0 \\ \hat{C}_4 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \hat{x} \\ \hat{i}_{\tilde{\mathcal{L}}} \\ \hat{u}_{\tilde{\mathcal{C}}} \\ \hat{u}_{\tilde{\mathcal{R}}} \end{bmatrix}$$

that approximate the original nonlinear system of model equations.

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MATLAB-Toolbox: PABTEC

Figure 1467.00464: PABTEC

General Simulation Reduction

Current directory: /circuit/pabtec/examples/05_NonLin

t_start: 0 sec
t_end: 7 sec

load matrices from m...
 generate matrices fro...

Computation

simulate original...
 Model Order Re...
 simulate reduce...

Original sys...
delta t: 0 sec
blaford: 7

Reduced sys...
delta t: 0.01 sec
blaford: 3

Export

Save matrices in "...
 Save matrices in "...
 Export graphics

General information

Check topology

Parameters for the reduction

reduction: 1
red tol: 1e-4
red out: -1
red. iout: 3

Parameters for the (inner) LR-ADI iteration

maxit: 4
minres: 1e-06

ADI parameters

with_rs: 5
miniu: 0
iout: 3
prol: 3

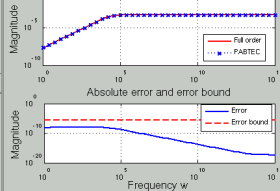
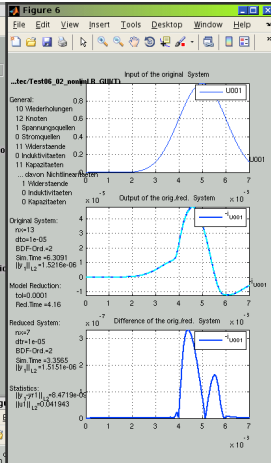
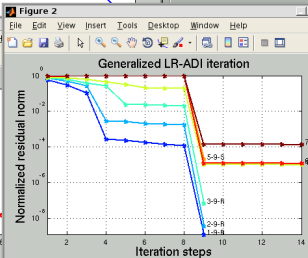
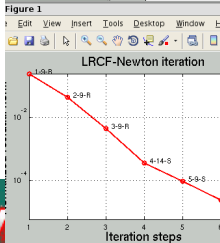
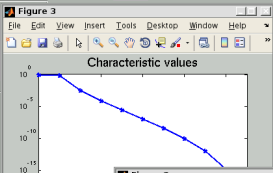
kp: 20
kl: 20
ks: 22
meth: h
iout: 1

Parameters for the (outer) Newton iteration

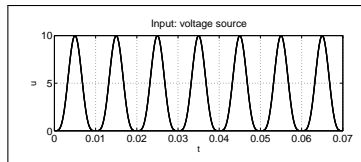
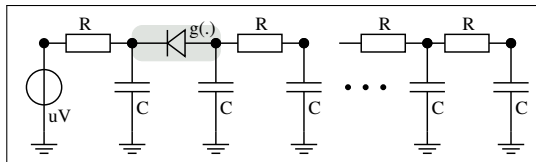
maxit: 15
minres: 1e-05

Plots

ADI (+Newton) absolute error...
 characteristic v... frequenc resp...
 relative error abs. error/-bo...
 positive realness freq. reson...



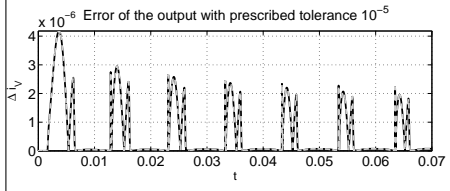
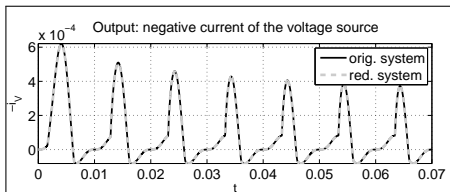
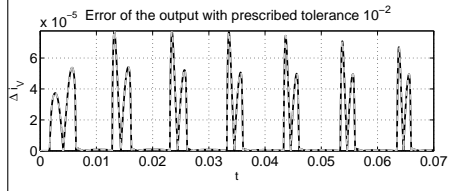
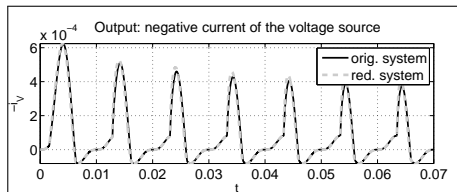
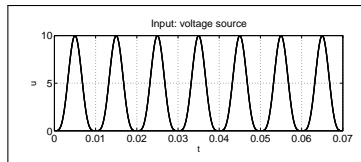
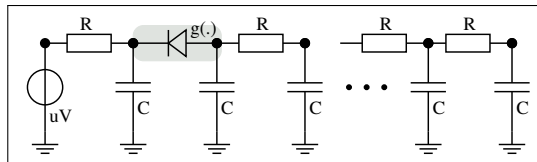
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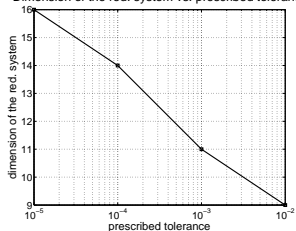
502	nodes	1	voltage source
501	linear capacities	1	output
500	linear resistors	1	diode

State dimension of the model equations $n = 503$

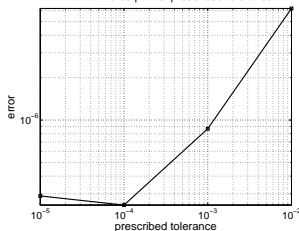
Simulation is done for $t \in [0s, 0.07s]$ using BDF method of order 2 with fixed stepsize of length $1 \cdot 10^{-5}$. The computations are done with MATLAB.



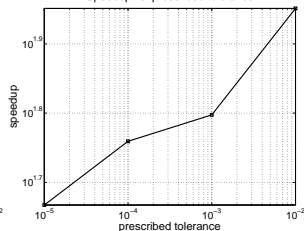
Dimension of the red. system vs. prescribed tolerance



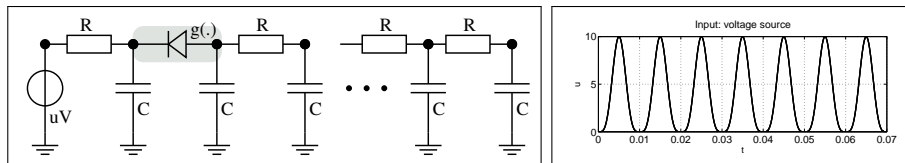
Error of the output vs. prescribed tolerance



Speedup vs. prescribed tolerance



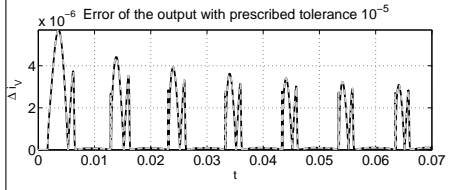
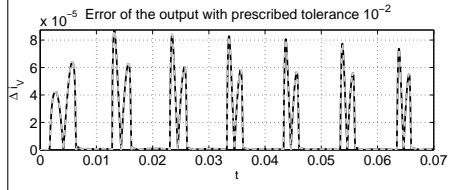
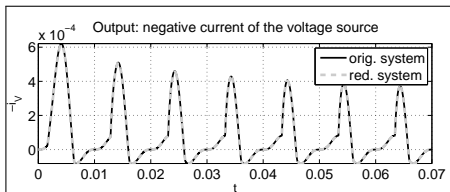
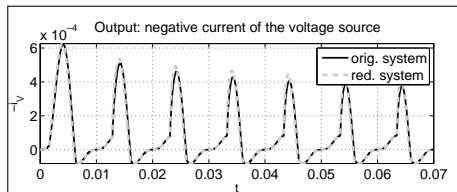
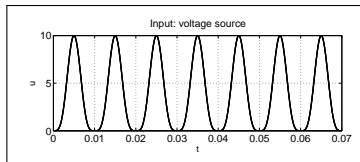
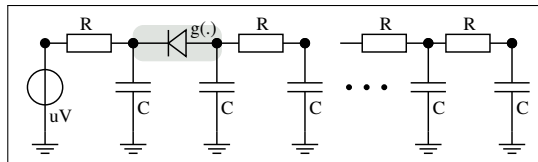
dimension of the original system	503	503	503	503
simulation time for the original system	6772s	6772s	6772s	6772s
prescribed tolerance for the model reduction	1e-02	1e-03	1e-04	1e-05
time for the model reduction	7s	8s	27s	46s
dimension of the reduced system	9	11	14	16
simulation time for the reduced system	76s	108s	118s	146s
obtained error of the output of the red. system	6.2e-06	8.7e-07	2.5e-07	2.9e-07
speedup	89.4	62.7	57.5	46.5



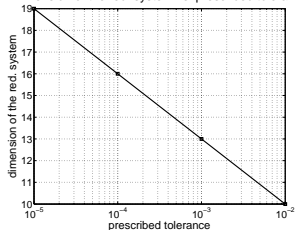
1502	nodes	1	voltage source
1501	linear capacities	1	output
1500	linear resistors	1	diode

State dimension of the model equations $n = 1503$

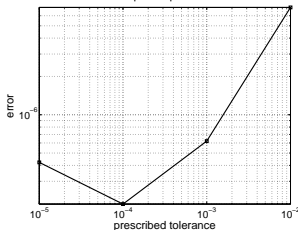
Simulation is done for $t \in [0s, 0.07s]$ using BDF method of order 2 with fixed stepsize of length $1 \cdot 10^{-5}$. The computations are done with MATLAB.



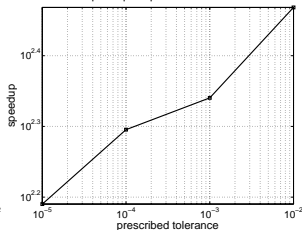
Dimension of the red. system vs. prescribed tolerance



Error of the output vs. prescribed tolerance

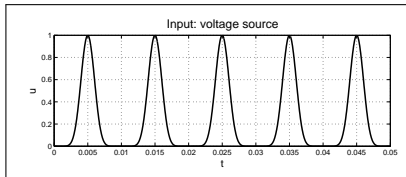
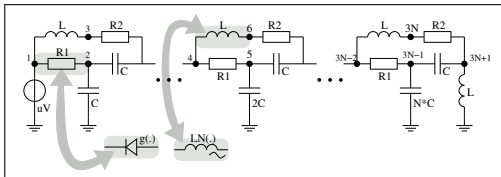


Speedup vs. prescribed tolerance



dimension of the original system	1503	1503	1503	1503
simulation time for the original system	24012s	24012s	24012s	24012s
prescribed tolerance for the model reduction	1e-02	1e-03	1e-04	1e-05
time for the model reduction	15s	24s	42s	61s
dimension of the reduced system	10	13	16	19
simulation time for the reduced system	82s	110s	122s	155s
obtained error of the output of the red. system	7.0e-06	6.2e-07	2.0e-07	4.2e-07
speedup	294.0	219.0	197.4	155.0

Example 02 - Problem

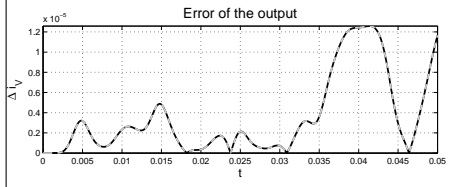
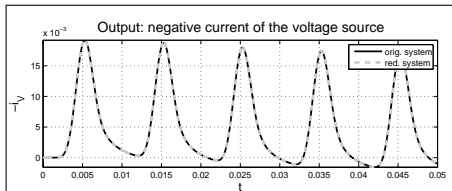
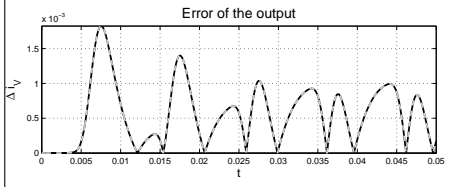
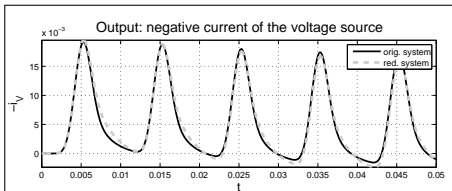
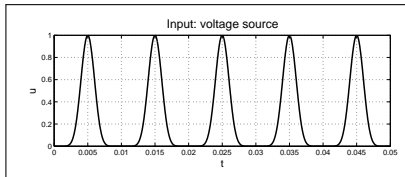
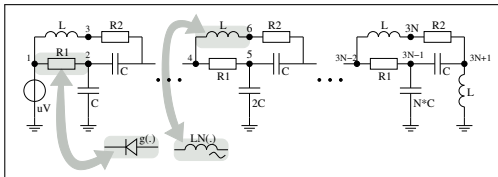


3001	nodes	1	voltage source
2000	linear capacities	1	output
1990	linear resistors	10	diode
991	linear inductors	10	nonlinear inductors

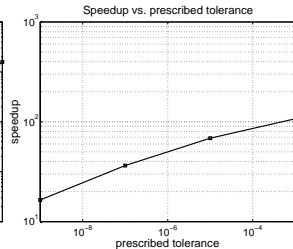
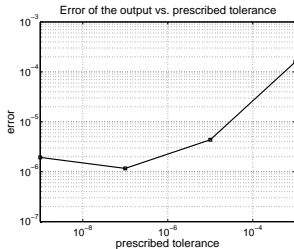
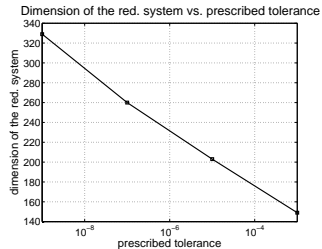
State dimension of the model equations $n = 4003$

Simulation is done for $t \in [0s, 0.05s]$ using BDF method of order 2 with fixed stepsize of length $5 \cdot 10^{-5}$. The computations are done with MATLAB.

Example 02 - Simulation results



Example 02 - Efficiency



dimension of the original system	4003	4003	4003	4003
simulation time for the original system	4557s	4557s	4557s	4557s
prescribed tolerance for the model reduction	1e-03	1e-05	1e-07	1e-09
time for the model reduction	902s	822s	834s	900s
dimension of the reduced system	149	203	260	329
simulation time for the reduced system	43s	67s	125s	277s
obtained error of the output of the red. system	1.6×10^{-4}	4.4×10^{-6}	1.16×10^{-6}	1.9×10^{-6}
speedup	107.2	68.5	36.5	16.5

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 - ▶ **Summary**

- ▶ We developed a model order reduction approach for the model equations of nonlinear circuits.

- ▶ We developed a model order reduction approach for the model equations of nonlinear circuits.
- ▶ The developed model reduction technique bases on ...
 - ▶ decoupling of linear and nonlinear subcircuits
 - ▶ model reduction of the remained linear part
 - ▶ recoupling of the reduced linear subcircuit with the unchanged nonlinear subcircuit

- ▶ We developed a model order reduction approach for the model equations of nonlinear circuits.
- ▶ The developed model reduction technique bases on ...
 - ▶ decoupling of linear and nonlinear subcircuits
 - ▶ model reduction of the remained linear part
 - ▶ recoupling of the reduced linear subcircuit with the unchanged nonlinear subcircuit
- ▶ The efficiency and applicability of the proposed model reduction approach was demonstrated on several numerical examples.

Workshop: MODRED 2010

MODEL REDUCTION FOR COMPLEX DYNAMICAL SYSTEMS



December 2-4, 2010
TU Berlin, Germany



Invited speakers: Michel S. Nakhla (Carleton University, Ottawa)
Joel R. Phillips (Cadence Berkeley Laboratories, San Jose)
Timo Reis (TU Hamburg-Harburg)

Important date: Registration: November 1, 2010

<http://www3.math.tu-berlin.de/modred2010>

Thank you for your attention.